

On the Measurement of the Resistivity in an Exploding Wire Experiment

Luis Bilbao, Gonzalo Rodríguez Prieto

Abstract—Explosion of a metallic wire due to a large electrical current can be used for studying metallic states difficult to reach with other methods. Due to experimental constraints, direct measurement of the voltage drop across the wire is impractical, although many characteristics of the metal state in the wire can be derived from these waveforms. Usually, the transformation of the electrical signals is made with the assumption of a lumped model for all the elements of the circuit, including the wire. We discuss the validity of a lumped model, and we show that due to the variation in time of the current density distribution on the wire, this model will not provide accurate values for the wire resistivity. Wire resistivity inaccuracies are specially clear in gas and plasma states, due to the diffusion and movement of the current that produce a large variation of the magnetic flux inside the wire.

In order to obtain more precise results in the resistivity of the wire metal, regardless of its state, a better approach is the use of the Faraday's law of induction on a path along the border of the wire. Our experiments of exploding wires in atmospheric air present the advantage of the clear electrical boundary between the expanding wire and the surrounding air, where no current circulates. As the state of the wire boundary layer changes from solid to plasma, it is possible to estimate the resistivity of the metal in those states in a more precise way.

Index Terms—Circuit analysis, metals, atmospheric-pressure plasmas, exploding wire, resistivity.

I. INTRODUCTION

WHEN a large electrical current passes through a metallic wire of the proper dimensions, typically 100 μm of diameter and centimeters length, the metallic wire is heated rapidly by Joule effect, becoming liquid, then gas, to later be transformed in plasma. This system is called exploding wire, and it is well known to science since a long time. It had been used in multiple endeavors, because the rich phenomena that can be accessed with it. Broad examples of the use of exploding wire are the general use as generator mechanism for blast waves [1] or the better understanding of the fuse dynamics through experiments like in the work of Vermij [2]. Exploding wire systems can also be used for important industrial or military applications, like in the preparation of metallic nano-powders reviewed by Kotov et al. [3], or the study of the mitigation of blast waves by foam, through the use of a surrogate setup, as in the recent work of Liverts et al. [4].

L. Bilbao is with Universidad de Buenos Aires, Facultad de Ciencias Exactas, Departamento de Física, Buenos Aires, Argentina and CONICET, Infip, Buenos Aires, Argentina.

G. Rodríguez Prieto is with Universidad de Castilla-La Mancha, Spain and INEL, Ciudad Real, Spain

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In order to create the high current necessary for this phenomenon, large capacitors and high voltages are necessary in the electrical circuit that delivers the current to the wire load. Main circuit characteristics, like the total inductance, resistance and capacitance due to the capacitor bank, can be well modeled by a RLC model, while the spark-gap and the wire need a different model.

Experimental voltage waveforms are usually obtained from probes attached to the circuit. Due to experimental constraints, sometimes the voltage probes are not exactly placed between the wire extremes, but separated by fixed circuit elements from them, for example, as described in the experimental works [5]–[9].

Additionally, there is a problem with the modeling of the experimentally obtained voltage and current signals, in the sense that the exploding wire electrical characteristics must be included in this modeling. In order to interpret experimental measures of current and voltage and, therefore, resistance or other electrical parameters, some authors use simple models assuming the wire as a lumped element. With this approach, the resistance of the wire, and in later stages, of the plasma, can be measured indirectly using the voltage signals, as in [10]–[13].

Despite the broad use of the lumped model for the wire, the description is not accurate during the time when the diffusion of the electrical current is important or when a large variation of the resistivity occurs within the wire. For example, Z-pinch system dynamical evolution, in which a cylindrical array of wires are made to implode radially by the gradient of the self-magnetic field of the current flowing through the wire array, cannot be described by just a lumped model for every wire [14], [15]. In fact, it was already noticed that in such systems, the plasma spatial and temporal distribution around every wire following the first moments of the electrical discharge is important in the dynamical description of the Z-pinch [16], [17].

So, a description of the exploding wire circuit, accounting for the distributed nature of the phenomenon, is important to understand and better qualify the wire explosion by means of an intense electrical current. Therefore, we present here a different approach to the description of the circuit. It is based on the derivation of the circuit equation by means of the Faraday's law of induction.

II. ELECTRICAL CIRCUIT

The experimental setup of ALEX (ALambre EXplosivo, exploding wire acronym in Spanish), the exploding wire experiment motivating this work, including an optical streak camera,

has been described in a previous work [18]. Nevertheless, it is worthwhile to recall here a description of its electrical circuit and the associated probes. The capacitor bank, two parallel Castor oil capacitors of $1.1 \mu\text{F}$ each, is charged by a high voltage source of a maximum voltage of 65 kV. Upon arrival to the desired nominal voltage, the high voltage source is disconnected from the circuit, and the spark-gap is triggered (see fig. 1). Later on, when the spark-gap switch is fully closed, the circuit is equivalent to a RLC circuit plus the wire load. Wire was made of Copper, with a length of 31 mm and diameter of $50 \mu\text{m}$.

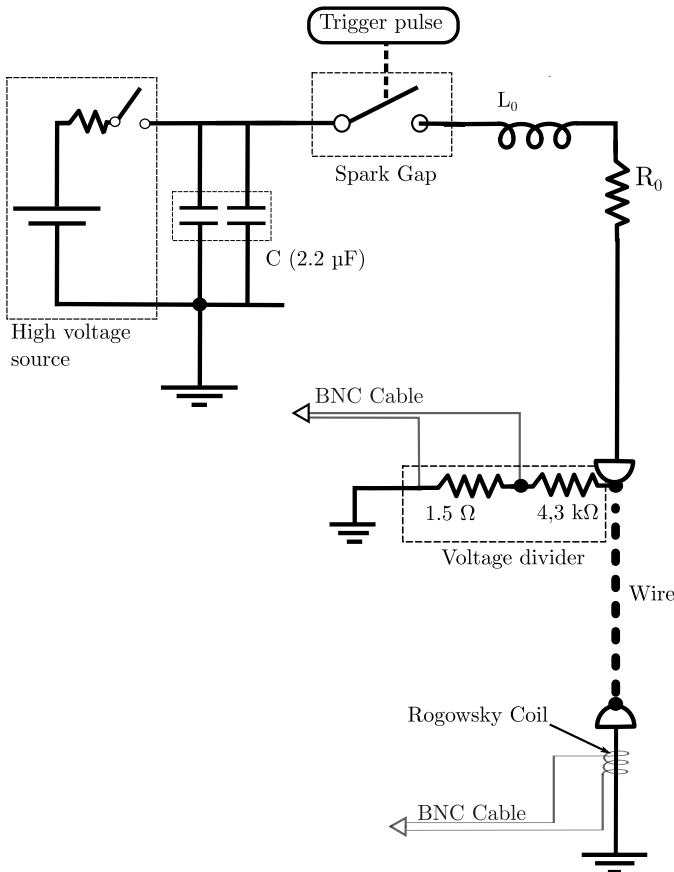


Fig. 1. ALEX electrical scheme. R_0 , L_0 and C represent the lumped circuit resistance, inductance and capacitance, respectively. Open arrows indicate the BNC connections to the oscilloscope.

The use of lumped element model for the exploding wire is not always the best approach. The discharge on the wire produces a current that diffuses due to its resistivity. During this stage, it is impossible to define separately a resistance and an inductance as lumped parameters of the circuit [19]. Notwithstanding this fact, lumped element models for the wire have been used for more than 50 years [6], [20]–[22], and, therefore, it is a common practice to refer to resistance and inductance of the wire. The practical use of such approach is justified as a way for obtaining the resistivity of the metal as a function of time from the electrical signals and the wire/plasma radius evolution. An example, which clearly states this lumped model for the wire, is the work of Sasaki et al. [10]. Be aware that the above model is not of general application because it

requires a homogeneous evolution in density and resistivity.

The above hypothesis is not usually well justified in all the stages of the exploding wire evolution. For example, just after the initial rise of electrical current, its value drops almost to zero while the full voltage is across the wire. This low current stage, called dark pause, lasts for a given time until a second surge of the current develops, see fig. 2, and under appropriate conditions it can last several μs .

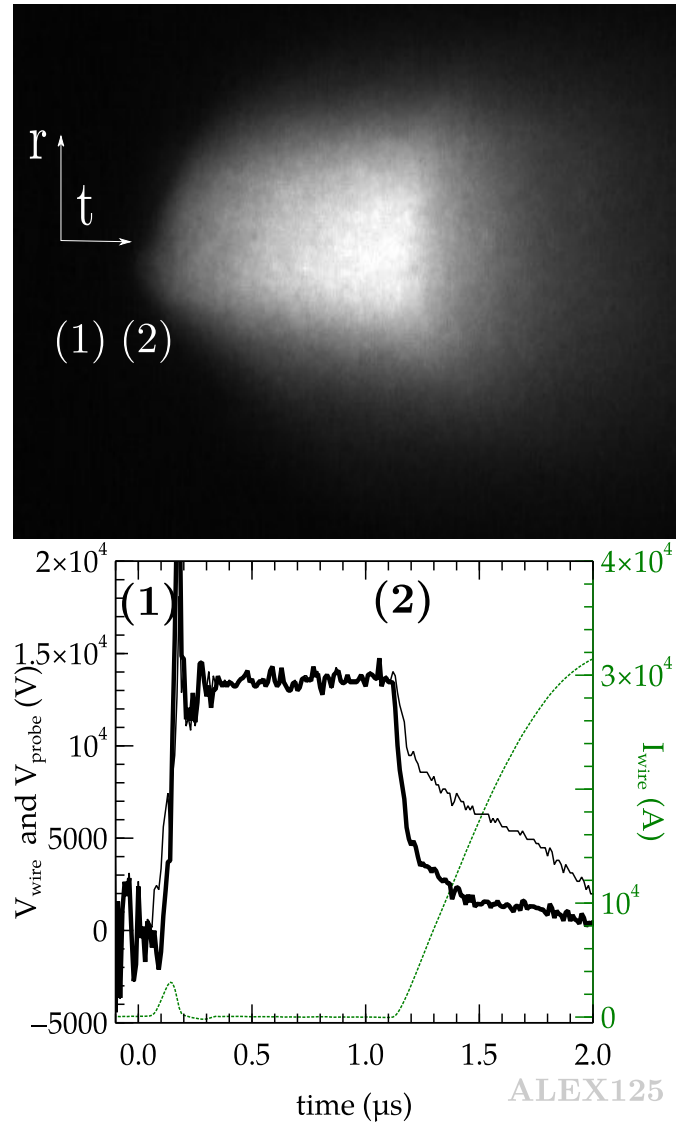


Fig. 2. (Color Online) In the upper panel, exploding wire streak image is shown. Time in the horizontal axis is $20 \mu\text{s}$, and space in the vertical axis, 24 mm . Bottom panel shows the voltage through the wire (—), the signal of the voltage probe (—), and the current (···). Note that current during the dark pause is different from zero. Numbers in both panels corresponds approximately to the same moments in both the streak image and the signals. Charging voltage was 15 kV with a wire diameter of $50 \mu\text{m}$.

The elapsed time between the first and the second surges of the current will depend on the resistivity of the wire and the distribution of the current. The wire is heated by Joule effect, and from the energy given to the wire system, we find that after the first current surge the wire melts and later on, its external layer starts to vaporize. When part of the

132 wire vaporizes, the sudden increase of the wire resistivity
 133 produces a sharp drop in the current. Although small, a
 134 residual current continues flowing and heating up the wire and
 135 a vaporization wave progresses through the wire. This means
 136 that different states (liquid, liquid/gas, and gas) may coexist
 137 in the wire, having very large difference in their resistivities.
 138 Indeed, the experimental parameters were chosen to allow
 139 for the exploration of classical states of matter in thermal
 140 equilibrium, in sharp contrast with more usual exploding wire
 141 experiments, aimed to explore warm dense matter conditions.

142 The use of one phase homogeneous model like the used
 143 in [10] is inadequate in our experimental conditions, where
 144 the dark pause stage is characterized by a large current
 145 diffusion at the beginning and the end, in addition to a two
 146 different homogeneous phases, an inner liquid surrounded by
 147 an expanding gas, during the remaining time of quasi constant
 148 current. Thus, a lumped element model of this stage will not be
 149 of sufficient accuracy to evaluate the Joule heating contribution
 150 to metallic gas. In order to increase the accuracy and allow for
 151 a better quantitative estimation of the gas resistivity, a model
 152 for the electrical circuit, based on Faraday's law of induction,
 153 that considers the wire as an extended entity is described in
 154 the next section.

155 III. CIRCUIT EQUATION

156 In order to model the electrical circuit, it is divided into three
 157 distinct parts: a) capacitors, cables, electrodes and connections,
 158 b) the spark-gap, and c) the exploding wire.

159 Part a) can be modeled with a RLC lumped element model,
 160 as it is experimentally seen using a short circuit. In our case,
 161 the wire is removed and the cathode is displaced until it
 162 touches the anode, i.e. no short circuit element is added.

163 On the one hand, the measured current derivative, dI/dt ,
 164 can be perfectly fitted by a damped cosine at the later stage
 165 where the voltage drop across the spark-gap is zero. The values
 166 of the lumped elements R_0 and L_0 are obtained from the fit
 167 with high precision, see fig. 3 (the capacitance, C , is provided
 168 by the manufacturer). The difference between the fitted RLC
 169 signal and the actual signal that is seen at the beginning of the
 170 signal (the first μs in fig. 3) is an oscillation due to the finite
 171 closure time of the spark-gap.

172 On the other hand, the measured voltage (i.e. the voltage
 173 drop in the cathode, when no wire is present) is perfectly fitted
 174 by a lumped resistance, $R_{cathode}$, and inductance, $L_{cathode}$.
 175 These values (different from R_0 and L_0) are obtained using a
 176 multiple linear regression of V_{probe} on dI/dt and the current
 177 I (obtained by numerically integrating dI/dt) by noting that,
 178 under short circuit conditions,

$$179 \quad V_{probe} = R_{cathode}I - L_{cathode} \frac{dI}{dt}. \quad (1)$$

180 The fitted signal of the voltage probe and its residual are also
 181 plotted in fig. 3.

182 Due to the fact that the closure time of the spark-gap is
 183 much shorter than the period of the discharge, the behavior of
 184 the spark-gap, part b), is well modeled by a variable voltage

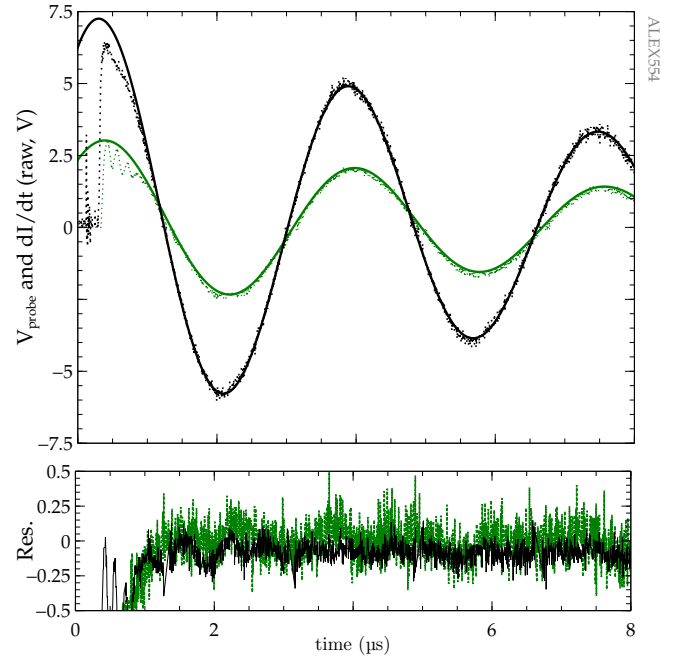


Fig. 3. (Color Online) ALEX short circuit current derivative and voltage raw waveforms and their fit (top panel) with the residual (down panel) when charged at 10 kV. Rogowski probe voltage experimental (\cdots) and fit (—) values are black lines, meanwhile voltage divider experimental (\cdots) and fit (—) traces are depicted in green in both panels. Note the large value of the residual at the beginning of the discharge, when an oscillation is produced by the finite closure time of the spark-gap.

185 drop. For the voltage variation across the spark-gap, $V_{sg}(t)$,
 186 we have used [23]

$$187 \quad V_{sg}(t) = \frac{2V_0}{1 + \exp(t/\tau)}, \quad (2)$$

188 where V_0 is the initial charging voltage of the capacitor, τ
 189 the spark-gap closure time (30 ns in our case), and $t = 0$
 190 corresponds to the trigger of the spark-gap.

191 To model the exploding wire, part c), we start by assuming
 192 that the current is not homogeneous across the wire section,
 193 that is, the current is distributed in space, and therefore flows
 194 through different paths between the electrodes.

195 Usually Kirchhoff's circuit laws are used to solve the current
 196 and voltage in an exploding wire experiments, regardless that
 197 it is only valid for a lumped element model. When the current
 198 is spatially distributed a more precise way to write the circuit
 199 equation is by means of the Faraday's law of induction, along
 200 a closed path across the full circuit, that goes through the
 201 lumped elements, the spark-gap, and the exploding wire as

$$202 \quad \frac{d\Phi}{dt} = - \oint \mathbf{E}' \cdot d\mathbf{l}, \quad (3)$$

203 where Φ is the magnetic flux enclosed by the path, and \mathbf{E}' the
 204 electric field in a system fixed to the path ($d\mathbf{l}$). The path may
 205 be a material or an immaterial one. Also, it may be a fixed
 206 or a mobile path relative to the lab system. In any case the
 207 electric field \mathbf{E}' is evaluated in a system fixed to the path, not
 208 to the lab.

209 Along the part modeled with the lumped element model, the

path is unique (i.e., lumped elements have no thickness), while on the exploding wire (whether on its initial state or during its evolution) any path that connects the electrodes may be used.

The integration of the electric field on the part that is modeled by a lumped element is straight forward, because it holds that

$$\int_{\text{inductance}} \mathbf{E}' \cdot d\mathbf{l} = L_0 \frac{dI}{dt}, \quad (4)$$

$$\int_{\text{resistance}} \mathbf{E}' \cdot d\mathbf{l} = R_0 I, \quad (5)$$

and

$$\int_{\text{capacitor}} \mathbf{E}' \cdot d\mathbf{l} = -\frac{Q}{C}, \quad (6)$$

where Q is the charge.

On the other hand, as it was mentioned above, the spark-gap can be modeled by a variable voltage, that is

$$\int_{\text{spark-gap}} \mathbf{E}' \cdot d\mathbf{l} = V_{sg}(t). \quad (7)$$

Finally, for the exploding wire it is necessary to choose a path that connects the electrodes. For example, using path a as shown in fig. 4, left panel, (3) becomes

$$\frac{d\Phi_a}{dt} + \int_a \mathbf{E}' \cdot d\mathbf{l} = \frac{Q}{C} - R_0 I - L_0 \frac{dI}{dt} - V_{sg}(t), \quad (8)$$

where Φ_a is the magnetic flux enclosed by the whole circuit that is closed by the path a in the wire. Any other path may be used as well, including a mobile path along the wire boundary (see fig. 4, central panel), in which case we have

$$\frac{d\Phi_b}{dt} + \int_b \mathbf{E}' \cdot d\mathbf{l} = \frac{Q}{C} - R_0 I - L_0 \frac{dI}{dt} - V_{sg}(t). \quad (9)$$

Clearly, (8) and (9) produce the same voltage drop in the circuit, since Faraday's law along a closed path formed by a and b , gives

$$\frac{d\Phi_a}{dt} + \int_a \mathbf{E}' \cdot d\mathbf{l} = \frac{d\Phi_b}{dt} + \int_b \mathbf{E}' \cdot d\mathbf{l}. \quad (10)$$

(see fig. 4, right panel).

The above relationship also shows an important fact when diffusion of the current is the dominating phenomenon: it is impossible to meaningfully define separately a resistance and an inductance of the wire as lumped parameters in the circuit [19]. This can be better seen using the Ohm law in a system fixed to the wire material, where the electric field is evaluated, that is

$$\mathbf{E}' = \rho \mathbf{j}, \quad (11)$$

where ρ is the resistivity and \mathbf{j} the current density, then (10) becomes

$$\frac{d\Phi_a}{dt} + \int_a \rho \mathbf{j} \cdot d\mathbf{l} = \frac{d\Phi_b}{dt} + \int_b \rho \mathbf{j} \cdot d\mathbf{l}. \quad (12)$$

Equation (12) shows that the inductive (first terms in each side) and the resistive parts (second terms in each side) depend on the path, thus a unique definition of the total inductance and resistance of the wire can not be made unless a uniform current density distribution is present. The assumption of an uniform

radial distribution of the current in the wire cross-sectional area is not valid when the diffusion time of the magnetic field of the current is shorter than the typical time scale of the process. Also, there is another fact that prevents from having an uniform current density distribution even in the long time scale. In our experiments a liquid core coexists with a surrounding metallic gas. The large difference in the resistivity of both phases produces a large difference in the current density therefore, the radial distribution is not uniform in the long time scale (microsecond in our experiments). Therefore, we are not allowed to use the hypothesis of current density uniformity in these calculations.

Although (8) and (9) can be used interchangeably, we have chosen the path b along the border of the wire. Assuming cylindrical symmetry, the azimuthal magnetic field B outside the border of the wire is

$$B = \frac{\mu_0 I}{2\pi r_b}, \quad (13)$$

where r_b is the radius of the border and I the total current circulating through the wire, that is also related to the integration of the current density through the section ($d\mathbf{S}$) of the wire as

$$I = \int \mathbf{j} \cdot d\mathbf{S}. \quad (14)$$

Note that the electrical current does not circulate beyond the wire border because of the air surrounding the wire. Thus, the calculation of the magnetic flux enclosed by a path along the border of the wire and the returning plate, can be simply calculated as

$$\Phi_b = L_b I,$$

being L_b a geometrical relationship equivalent to an inductance. In our device, the returning electrical path is a conducting plate separated by a distance d from the axis of the wire, in which case L_b can be approximated by

$$L_b = \frac{\mu_0 l}{2\pi} \cosh^{-1} \left(\frac{d}{r_b} \right), \quad (15)$$

where l is the length of the wire. Therefore, (9) becomes

$$\frac{d(L_b I)}{dt} + \int_b \rho \mathbf{j} \cdot d\mathbf{l} = \frac{Q}{C} - R_0 I - L_0 \frac{dI}{dt} - V_{sg}(t), \quad (16)$$

Note that L_b cannot be considered as the wire inductance, since further magnetic flux that varies with time is inside the wire. Similarly, the second term of the left hand side of (16) cannot be replaced by a lumped resistance voltage drop in the form $R_b I$. Making such substitution implies that the terms R_b and dL_b/dt will be undistinguished between them in the electrical signals, thus a unique solution for R_b can not be experimentally obtained. This fact has been pointed out by Fridman [24] in the sense that time evolution of the resistance and the inductance obtained from the oscillograms of current and voltage is not a single-valued problem.

Actually, there is no need of calculating a wire inductance or resistance in order to solve the circuit equation. Instead, a "boundary inductance" (L_b) and a "boundary resistive voltage drop" ($\int_b \rho \mathbf{j} \cdot d\mathbf{l}$) are sufficient. Inside the wire, the Faraday's law of induction may be further used for deriving its structure,

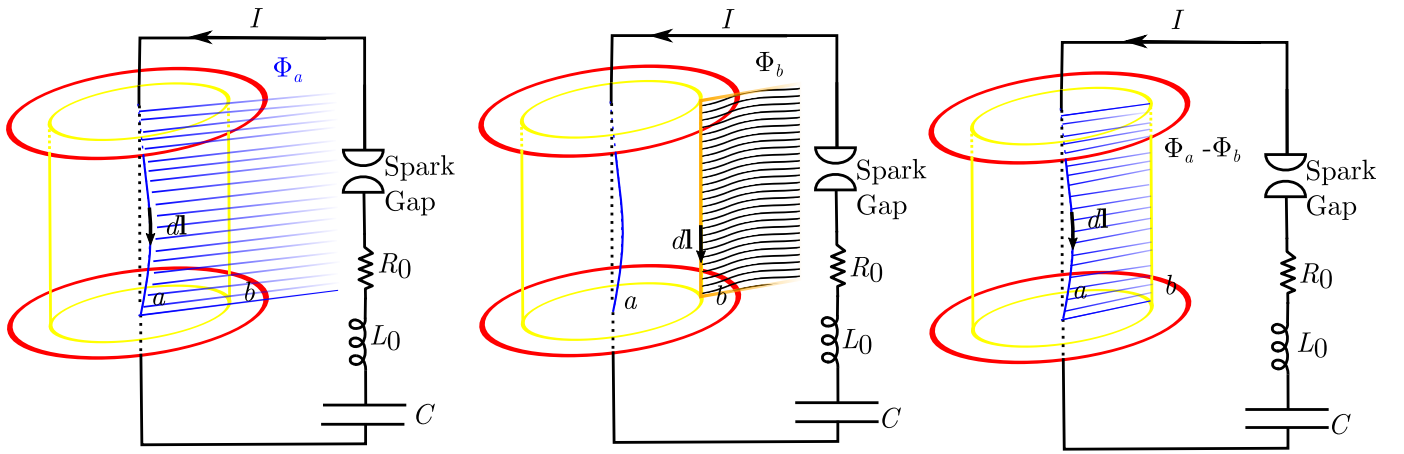


Fig. 4. (Color Online) ALEX integration paths with a schematic representation of the magnetic flux through the circuit produced by the wire explosion. Left panel indicates the path through the center of the plasma, central panel through the outer border of the plasma, and in the right panel, the flux difference between previous paths.

as in (12).

The set of equations (16) and (15) plus the exploding wire evolution, together with the initial conditions

$$Q(t=0) = CV_0; \quad I(t=0) = 0; \quad \frac{dI}{dt}(t=0) = 0$$

solves the circuit.

From an experimental point of view, it is possible to obtain information on the resistivity, from the voltage drop on the wire, left hand side of (16), which is:

$$V_{wire} = \frac{d(L_b I)}{dt} + \int_b \rho_j \cdot dl. \quad (17)$$

Under cylindrical symmetry, i.e. assuming no axial dependence, the above becomes

$$V_{wire} = \frac{d(L_b I)}{dt} + \rho_b j_b l, \quad (18)$$

where ρ_b and j_b are the resistivity and current density in the border of the wire, respectively.

From (18) the resistivity of the boundary layer can be obtained from the electrical signals as long as the current density in the border, and the movement of the boundary layer are also known.

In our experiment, for the measurement of the voltage drop on the wire, we have to take into account that the position of the voltage divider is at the connection between the anode and the wire, as it was previously mentioned. Therefore, the measured voltage drop, V_{probe} , is the sum of the voltage drop across the exploding wire, V_{wire} , plus the voltage drop in the cathode and connections, therefore

$$V_{wire} = V_{probe} - R_{cathode} I - L_{cathode} \frac{dI}{dt}, \quad (19)$$

where the current I is obtained by numerically integrating dI/dt , which is measured with the Rogowski coil.

IV. NUMERICAL SIMULATION

The aim of the numerical simulation is to help to understand the influence of the magnetic flux variation and the nonuniform

resistivity inside the wire in the interpretation of electrical signals.

During the dark pause the drop of the current indicates an important increment of the resistivity. The variation of resistivity of the copper until the boiling temperature [25] is not high enough to explain the observed drop, suggesting that the wire has been partially vaporized, because the energy provided to the wire up to this time, is not enough to vaporize the whole wire.

The end of the dark pause occurs when the metallic gas is ionized and the current is re-establish through the wire. Note that for simplicity we call "wire" to any state, that is solid, liquid, gas or plasma.

In order to show this idea, a 1D numerical code has been used to simulate the exploding wire dynamics. A key hypothesis in this approach is the symmetry of the wire evolution during the time of interest, that is the dark pause in the present work. To the purpose of the present analysis, the symmetry of the wire observed during the dark pause by means of streak and framing pictures, allows us the use of a 1D code where magnitudes depend only on the radial coordinate. Therefore, we assumed cylindrical symmetry with only radial dependence. In this way the plasma expansion can be approximate by a 1D system where the spatial coordinate corresponds to the radius.

We have adapted a previous developed 3D code [26], that has been used to simulate different physical problems such as double-base chemical propellant combustion, ignition and propagation of a thermonuclear detonation wave, and, the development of the Kelvin-Helmholtz (KH) instability in the magnetopause. In the present version, a Mie-Grüneisen equation of state for solid and liquid was used, and the ionization state was obtained from a Saha equation. This is justified in the fact that our focus is the study of the dark pause, that ends when a cold plasma is formed. The successive evolution of the plasma at higher temperatures is out of the scope of this study. The code solves the equations of continuity, momentum, and energy plus Maxwell's equations in the wire, coupled to the circuit equation (16) with conditions (13) and (14). The

376 electrical resistivity of copper is readily available for solid,
377 liquid [25] and plasma states, but not for the gas state.

378 Pressure and density of the gas state are obtained from the
379 radial expansion in time, observed with the streak camera.
380 Estimating the pressure with the Rankine–Hugoniot relation-
381 ship for the shock into the air at atmospheric pressure and
382 ambient temperature gives values between 10 and 100 atm.
383 In combination with direct measurements of the radius of the
384 expanded gas, a fairly constant density on the order of 5×10^{18}
385 cm^{-3} is estimated. Further, due to the absence of radiation
386 coming from it, we can conclude that the metallic gas is in
387 classical neutral state which resistivity is expected to depend
388 on the gas temperature.

389 As a first approach we have used an ad hoc linear variation
390 of the resistivity, ρ_{gas} , with the temperature as:

$$391 \quad \rho_{gas} = \rho_0[1 + \alpha(T - T_{boil})], \quad (20)$$

392 being T the gas temperature in Kelvin, and $T_{boil} = 2940$ K
393 the boiling temperature of copper.

394 As an starting point for the determination of the parameters,
395 ρ_0 and α , an order of magnitude value was experimentally
396 obtained from the “boundary resistive voltage drop” mentioned
397 above. Then, using the code, the parameters were iterated
398 until a fairly good reproduction of the measured voltages and
399 currents through the wire were obtained. It has been observed
400 that the mean value of ρ_{gas} is related to the dark pause
401 duration, while the slope accounts for the variations of the
402 current and voltage in time.

403 This procedure was repeated for various initial charging
404 voltage values. The best estimates, using this rough method,
405 were $\rho_0 = 4 \times 10^{-3} \Omega \cdot \text{m}$ and $\alpha = 0.00045/\text{K}$.

406 The above estimates are intended only to illustrate the
407 differences that may arise when a lumped model is used for
408 the exploding wire. They cannot be taken as a precise value
409 of the copper gas resistivity.

410 The simulated electrical signals were obtained by solving
411 the numerical code coupled to the circuit equation, using the
412 values of our experiment, $C = 2.2 \mu\text{F}$, $L_0 = 142 \text{ nH}$, $R_0 = 5$
413 $\text{m}\Omega$, $V_0 = 15 \text{ kV}$, copper wire $50 \mu\text{m}$ in diameter and 31 mm
414 long. Fig. 5 shows the voltage drop in the wire (simulated and
415 measured) as a function of time, with the state of the outer
416 layer of the wire over-imposed, as it evolves from solid to
417 plasma. We observed (not shown in the figure) that the state
418 in the inner part of the wire differs from that of the external
419 layer. This means, that different states coexist in concentric
420 layers that evolve with different time scales.

421 V. MEASUREMENT OF THE RESISTIVITY

422 As it was mentioned in the Introduction, several authors
423 obtain a value of resistance, R , from the electrical signal, by
424 subtracting the “inductive” part from the voltage drop on the
425 exploding wire as

$$426 \quad R = \frac{V_{wire} - \frac{d(L_b I)}{dt}}{I}, \quad (21)$$

427 From (21) a mean resistivity may be inferred as

$$428 \quad \langle \rho \rangle = \frac{RS}{l}, \quad (22)$$

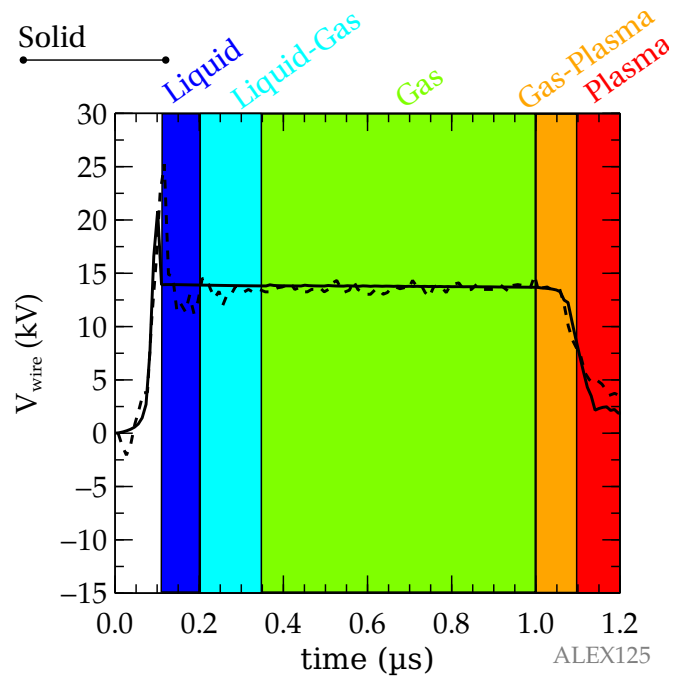


Fig. 5. (Color online) ALEX voltage waveforms (calculated: full line, measured: dashed line) with the states of the outer shell of the wire clearly indicated, for a copper wire with a diameter of $50 \mu\text{m}$ and with capacitors charged at 15 kV.

where l is the length and S the section of the wire. 429

Comparing with (18) and assuming cylindrical symmetry it follows that 430

$$431 \quad \langle \rho \rangle = \rho_b \frac{j_b}{\langle j \rangle}, \quad (23) \quad 432$$

where $\langle j \rangle = I/S$ is the mean current density. Clearly, when 433
no magnetic flux variation is inside the wire (this implies that 434
there is no current diffusion, either), the above relationship 435
gives a reasonable mean value since $\rho j = \text{const}$ along the 436
radius, unless large variation of the resistivity occurs inside 437
the wire (for example, when different states coexist), in which 438
case, a mean value has no significance. 439

Using the numerical code, a mean “measured” resistivity, 440
(22), was calculated and compared to a spatial mean resistivity, 441
 $\bar{\rho}$, over the radial coordinate, defined as 442

$$443 \quad \bar{\rho} = \frac{1}{r_b} \int_0^{r_b} \rho(r) dr. \quad (24)$$

Other mean values may be defined as well, but the aim of 444
this paper is to show the difficulty in interpreting (22) as a 445
representative value. 446

The percentage difference $(\langle \rho \rangle - \bar{\rho})/\bar{\rho}$ is plotted in fig. 6. 447
As can be seen, the difference varies up to $\pm 100\%$. From this 448
result it is clear that the resistivity experimentally obtained 449
from (21) and (22) may considerably depart from the actual 450
value. In the present example it is due to two main factors: 451
a) the diffusion of the current, and b) the different states that 452
simultaneously coexists in the wire. 453

The effect a) is clearly seen when the current varies in a 454
characteristic time smaller or similar to the diffusion time, 455
as it happens at the beginning and at the end of the dark 456

457 pause. At the beginning of the electrical discharge, the current
 458 density reaches its largest value in the wire surface due to the
 459 current concentration in this region, and diffuses to the center
 460 of the wire. When the current diffuses from the boundary to
 461 the center, the current density is maximum at the border (i.e.
 462 $j_b > \langle j \rangle$) producing an overvalued mean resistivity, $\langle \rho \rangle$, as
 463 can be seen from (23) and is shown in the initial times in
 464 fig. 6. The opposite happens when the current diffuses from
 465 the center to the boundary, or similarly when the magnetic
 466 flux decreases inside the wire (for example, due to a sudden
 467 expansion).

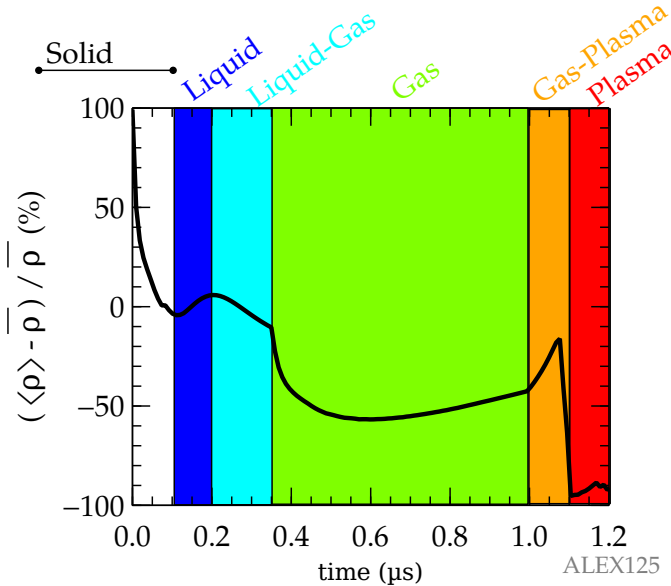


Fig. 6. (Color Online) Calculated percentage resistivity difference between (22) and (24). A copper wire of $50 \mu\text{m}$ diameter and 30 mm length, charged at 15 kV was used in the simulation. Background color correspond to the outer layer state, as in the previous figure.

468 As it was mentioned above, in the example of fig. 6,
 469 simulations have shown that the gas state of the outer layer
 470 coexists with the liquid state of the central layer (not shown
 471 in the figure). The coexistence of different states that evolve
 472 during the dark pause, produce a large difference of the
 473 resistivity (many orders of magnitude) across the wire making
 474 meaningless the concept of a mean resistivity such as (22).

475 VI. CONCLUSIONS

476 Calculation of resistivity from the experimental measure-
 477 ments of the electrical signals and the wire evolution, assuming
 478 a lumped element model as in (21) and (22), may give
 479 large difference relative to the actual resistivity, in particular,
 480 when the current varies during a time comparable to the
 481 diffusion time, or when different states coexist inside the wire.
 482 Therefore, the use of lumped element model for the exploding
 483 wire is at best an approximation, that may be use with care.

484 If the study of metal properties as a function of temperature
 485 and density is seek, a better strategy would be to obtain the
 486 resistivity of the outer layer of the wire by measuring the
 487 “boundary resistive voltage drop”. The external part of the
 488 wire varies its state from solid to plasma, thus, in principle,

the time variation of the resistivity corresponding to different
 states may be studied. The resistivity of the outer layer can be
 estimated, as long as the current density is known.

There is no need to estimate the resistivity of the inner part,
 nor its mean value on the wire. It is enough to study the outer
 layer, and, if possible, by also measuring temperature, density
 and current density.

It is worth mention that the measurements here presented
 assume the existence of thermal equilibrium, a condition not
 always achieved in such a dynamic environment as it is the
 exploding wire. In absence of this local thermal equilibrium
 condition, the obtained resistivity cannot be understood as any
 constitutive property of the exploding wire matter, independ-
 ently of its phase, neutral gas or plasma. Nevertheless, the
 main objective of this work is the resistivity measurement
 method for local thermal equilibrium systems, not the deter-
 mination of non-equilibrium states properties.

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