

VARIABLE-STRUCTURE CONTROL WITH COMPLEMENTARITY-INPUTS FOR A LEAN-BURN IC ENGINE OF A SERIES HYBRID VEHICLE

N. Roqueiro, E. Fossas, A.A. Martins Oliveira, and P. Puleston

ABSTRACT

This paper presents a robust controller for an internal combustion (IC) engine, as the first stage of a project to develop a hybrid light urban vehicle, running on ethanol in lean burn. In particular, this work focuses on the design of a sliding mode control for an IC engine of a series hybrid power train. The controller must allow for optimal speed regulation and high fuel efficiency. To attain the latter, a complementary operation mode is proposed for the system inputs. Simulation results are presented and discussed showing the viability and advantages of the control strategy employed.

Key Words: Sliding mode control, IC engine, nonlinear models.

I. INTRODUCTION

The control of internal combustion engines remains an active area for the application of advanced control methods [1–5]. Nowadays, growing interest in hybrid powertrains offers new opportunities for pollution reduction and fuel economy [6–8]. A series hybrid powertrain appears to be the most promising architecture due to its applicability to small engines and the possibility of intermittent operation. In the current research project, this concept is being explored by a multi-institutional international team focusing on a small series hybrid powertrain for a light urban vehicle [9]. The engine of the hybrid power train will run on ethanol in lean burn, using electric cold start and warm up. It will drive a synchronous electric generator, whose voltage is rectified to feed a DC bus. It is well known that a better performance and a

simpler design of a controlled rectifier can be obtained when the engine speed remains close its nominal value. Therefore, the primary objectives of this work are speed control and fuel efficiency. The efficiency attained must reach at least the US-EPA greenhouse gas score (US EPA - Vehicle Environmental Scoring) of 8. Also, the overall emissions must be in compliance with the levels stipulated in California's Super Ultra Low Emission Vehicle, SULEV-II.

This paper reports the results of the first stage of the ongoing research. Specifically, it describes a sliding-mode controller with complementary-inputs designed to improve the speed regulation and fuel efficiency of a four-stroke engine with lean-burn operation. In this initial phase, the benchmark mean value engine model (MVEM) developed by Crossley and Cook [10] is used. The original model parameters are maintained in this analysis, without loss of generality. Later this will be adapted with the actual engine parameters, once the testing of the laboratory internal combustion engine has been completed. In this context, the aim of this work was to assess the suitability of the sliding mode approach to control IC engines, with fixed open throttle under lean burn conditions, acting only on the spark advance (θ) and relative air/fuel ratio (λ).

It has been well established that adequate lean burn conditions result in higher fuel conversion efficiencies [11]. This occurs because the combustion products have a lower specific heat at the lower combustion temperature leading to an increase in the effective polytropic coefficient during expansion. This results in a higher production of work during the expansion stroke. However,

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01 the fuel conversion efficiency decreases for excessively
02 lean mixtures, because of the increase in the pressure
03 fluctuations from cycle to cycle and the longer combus-
04 tion period. Therefore, there is a lean operation limit,
05 for which a small cycle to cycle variation is observed
06 and the fuel conversion efficiency reaches a maximum.
07 Consequently, the controller should be able to robustly
08 regulate the engine speed at the optimal nominal value,
09 while maintaining the mixture within the lean operation
10 limit to a feasible extent.

11 The paper is organized as follows. In Section II the
12 model of the IC engine in lean combustion operation is
13 introduced. Section III presents the design of the sliding
14 mode controller with complementary inputs. Section IV
15 reports the numerical results obtained through compre-
16 hensive simulations. Finally, in Section V, conclusions are
17 drawn and the plans for future work are outlined.

18 II. INTERNAL COMBUSTION 19 ENGINE MODELLING

23 Mean value engine models are dynamic models that
24 describe dynamic engine variable responses (or changes
25 of states) as mean values rather than instantaneous
26 crank-angle resolved values [12]. These models do not
27 describe the variations in the state variables over the
28 smallest engine time scales for each cylinder, but rather
29 the average changes in state that the engine experiences
30 as a whole. The engine benchmark MVEM developed by
31 Crossley and Cook [10] and implemented by MathWorks
32 in SIMULINK is then presented (the values given for the
33 model constants are those available in MathWorks).

34 2.1 Emptying and filling gas exchange model

35 The emptying and filling can be modelled as the
36 gas exchange between two large reservoirs, whose flow
37 is controlled by the area and discharge coefficient of the
38 connecting passage.

39 According to the ideal gas law, the dynamic
40 equation for the pressure p_i (kPa) of the engine intake
41 manifold, assuming constant temperature and volume, is

$$42 \dot{p}_i = \frac{RT_i}{V_i} (\dot{m}_{at} - \dot{m}_{ap}) \quad (1)$$

43 where R (kJ/kg-K) is the universal gas constant for the
44 fuel-air mixture, T_i (K) is the gas manifold temperature,
45 which is assumed to be isothermal and constant, and
46 V_i (m³) is the manifold volume, assumed to be constant.
47 \dot{m}_{at} (kg/s) and \dot{m}_{ap} (kg/s) are the mass flow rates of the air

crossing the throttle valve and the mass flow rate of the
mixture entering the cylinders, respectively.

The latter is given by

$$54 \dot{m}_{ap} = -0.366 + 0.08979 n p_i \quad (2)$$

$$55 - 0.0337 n p_i^2 + 0.0001 n^2 p_i \quad (2)$$

56 where n (rad/s) is the crankshaft rotation speed.

57 The former is modelled as the product of two func-
58 tions f and g :

$$59 \dot{m}_{at} = f(\alpha)g(p_i) \quad (3)$$

60 where

$$61 f(\alpha) = 2.821 - 0.05231\alpha \quad (4)$$

$$62 + 0.10299\alpha^2 - 0.00063\alpha^3 \quad (4)$$

63 and

$$64 g(p_i) = \frac{2}{p_a} \sqrt{p_i p_a - p_i^2} \quad (5)$$

65 where α is the throttle plate angle and p_a (101.3 kPa) is
66 the atmospheric pressure. In this work α is assumed to be
67 constant and equal to $\frac{\pi}{2}$ (rad). Thus this topology elimi-
68 nates the throttle as an actuator, allowing a reduction in
69 the cost and complexity of the control system.

70 2.2 Crankshaft rotation speed model

71 The dynamic equation for the crankshaft rotation
72 speed, n , results from a balance between the load torque
73 and the torque generated by the engine:

$$74 \dot{n} = \frac{1}{I} (T_e - T_L) \quad (6)$$

75 where I (kg-m²) is the crankshaft load inertia, T_e (N-m)
76 is the engine torque generated and T_L (N-m) is the
77 load torque. T_L is assumed to be as a measurable
78 time-dependent variable with a known bound.

79 The torque generated by the IC engine is repre-
80 sented by a polynomial fit of measured data for different
81 engine operating conditions. The equation for the engine
82 torque is:

$$83 T_e = -181.3 + 379.36m_{ap} + 21.91 \left(\frac{A}{F}\right) - 0.85 \left(\frac{A}{F}\right)^2$$

$$84 + 0.26\theta - 0.0028\theta^2 + 0.027n - 0.000107n^2 \quad (7)$$

$$85 + 0.00048n\theta + 2.55\theta m_{ap} - 0.05\theta^2 m_{ap}$$

86 where $\left(\frac{A}{F}\right)$ (dimensionless) is the air-fuel mass ratio,
87 θ (deg) is the spark advance and m_{ap} (g/s) is the mass

of the gas loaded into the cylinder during the intake stroke, which takes place in the first π radians of the crankshaft rotation of the four-stroke cycle. Thus, in the model, m_{ap} was obtained by integrating the air mass flow from the manifold and resetting the integrator at the end of each intake stroke. This can be closely approximated as follows:

$$m_{ap} = \frac{\dot{m}_{ap}\pi}{n} \quad (8)$$

On the other hand, the air–fuel ratio can be computed as follows:

$$\left(\frac{A}{F}\right) = \frac{\dot{m}_{ap}}{\dot{m}_f} \quad (9)$$

where \dot{m}_f (kg/s) is the mass flow rate of the fuel flowing into the cylinders, injected by the actuators also in the first π radians of the crankshaft rotation.

Finally, the air–fuel ratio can be directly related to the relative air–fuel ratio λ through a constant:

$$\lambda = \frac{\left(\frac{A}{F}\right)}{a_s} \quad (10)$$

where the constant a_s is the stoichiometric air–fuel mass ratio, herein assumed to be 14.6. Note that $\lambda = 1.0$ at stoichiometry, $\lambda < 1.0$ for rich fuel mixtures and $\lambda > 1.0$ for lean fuel mixtures.

2.3 Fuel conversion efficiency

To conclude this section, it is important to define the fuel conversion efficiency η_f . Considering the lowest heating value of the fuel to be 42,000 kJ/kg, this is given by

$$\eta_f = \frac{1}{BSFC \cdot 42000} = \frac{T_e n}{\dot{m}_f 42000} \quad (11)$$

where BSFC stands for brake specific fuel consumption.

The fuel conversion efficiency is not used for control purposes in this paper. Instead, it is considered as an engine performance indicator.

2.4 NOx formation

The nitrogen oxides (NOx) mole concentration of the exhaust gas leaving the exhaust manifold can be approximated as [12]

$$X_{NO}(\lambda, \theta) = X_{NO,o}(1 + \Delta\lambda + \Delta\theta) \quad (12)$$

where $X_{NO}(\lambda, \theta)$ is a mapping developed around a fixed point $X_{NO,o}(\lambda_o, \theta_o)$, obtained for representative constant values for the air–fuel ratio λ_o and spark advance θ_o . Functions $\Delta\lambda$ and $\Delta\theta$ take into account variations with respect to λ_o and θ_o and take the form:

$$\Delta\lambda = k_{\lambda,0} + k_{\lambda,1}(\lambda - \lambda_o) + k_{\lambda,2}(\lambda - \lambda_o)^2 \quad (13)$$

$$\Delta\theta = k_{\theta,1}(\theta - \theta_o) + k_{\theta,2}(\theta - \theta_o)^2 + k_{\theta,3}(\theta - \theta_o)^3 \quad (14)$$

where the coefficients $k_{\lambda,0}, k_{\lambda,1}, k_{\lambda,2}, k_{\theta,1}, k_{\theta,2}$ and $k_{\theta,3}$ are curve fitted to measurements.

To obtain the NOx mass flow rate the following approximation can be used.

$$\dot{m}_{NOx}(\lambda, \theta) = X_{NO}(t) \frac{M_{NOx}}{M_{eg}} \dot{m}_{ap}(t) \quad (15)$$

where M_{eg} and M_{NOx} are the molar mass of the exhaust gas and the NOx gas, respectively. Since only relative amounts of NOx emission are of interest, for the ratio $\frac{M_{NOx}}{M_{eg}}$ a value of 1 can be adopted. This simplified expression yields a sufficiently well approximated trend for the NOx concentrations [12].

Note that, similarly to the conversion efficiency, in this work the NOx mass flow rate is not used for control purposes, but only as a relevant engine performance indicator.

III. DEVELOPMENT OF THE SM CONTROLLER WITH COMPLEMENTARITY INPUTS

Sliding mode techniques have proved to be effective in several areas of automotive applications [13–16], due to their inherent robust features. In particular, this approach to the control is suitable for the problem

Table I. NOx model parameters.

$X_{NO,o}$	2162
$k_{\lambda,0}$	2.1
$k_{\lambda,1}$	1993
$k_{\lambda,2}$	−46550
$k_{\theta,1}$	−91.1
$k_{\theta,2}$	3.14
$k_{\theta,3}$	0.127
λ_o	1
θ_o	25

01 under consideration, given that the control system
02 must be robust in relation to the intrinsic model
03 uncertainties and external disturbances, while being
04 able to deal with the highly nonlinear nature of the
05 IC engine.

06 In summary, the system has two physical inputs,
07 that is, the ignition spark advance θ and the relative
08 air–fuel ratio λ , while the output is the engine speed n . It
09 should be noted that the engine is running with full open
10 throttle, so that the air mass flow rate is not controlled.
11 With these inputs at hand, the control objectives are to
12 maintain the engine speed at a nominal value and to con-
13 sume the minimum mass of fuel that allows the former
14 objective to be met.

3.1 Fuel consumption minimization: implementation of complementary inputs

19 The first stage towards the development of the con-
20 troller is to coordinate the two physical inputs, λ and θ ,
21 so that they operate as a single "complementary-inputs
22 actuator", where one of the inputs takes control while the
23 other remains constant, depending on the engine speed.

24 This particular actuator operation is proposed to
25 fulfil the second control objective (*i.e.* minimum fuel con-
26 sumption during speed regulation). In this regard, certain
27 considerations regarding the IC engine operation must
28 be taken into account. It can be established that, for the
29 engine under study, on operating at the leanest admissi-
30 ble mixture (which implies the maximum relative air–fuel
31 ratio λ_{max}) the fuel consumption is reduced. Thus, when-
32 ever possible, it is preferable to maintain constant the
33 input $\lambda = \lambda_{max}$ and counteract the variation in the load
34 torque by appropriately varying θ . This procedure will
35 be adequate within a certain load range, delimited by the
36 maximum spark advance angle, θ_{max} (note that at θ_{max}
37 the maximum torque engine is generated). Therefore, once
38 θ_{max} is reached, this input is kept constant at $\theta = \theta_{max}$
39 and higher load demands must be satisfied by manipulating
40 the other input λ .

41 The outcome of this analysis is that, in practice, fuel
42 consumption can be minimized via the operating inputs
43 θ and λ as follows:

$$0 \leq (\theta_{max} - \theta) \perp (\lambda_{max} - \lambda) \geq 0 \quad (16)$$

47 where the symbol \perp stands for complementarity, *i.e.*,
48 when $(\theta_{max} - \theta)$ is not zero, then $(\lambda_{max} - \lambda)$ is equal to
49 zero, and vice versa.

50 In accordance with the following algorithm:

$$52 \text{ Case : } \theta < \theta_{max} \rightarrow \text{Mode } \theta \begin{cases} \theta \text{ active input} \\ \lambda = \lambda_{max} \end{cases}$$

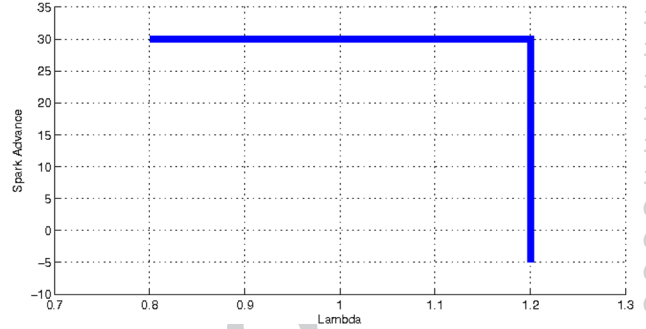


Fig. 1. Domain of complementary inputs.

$$54 \text{ Case : } \lambda < \lambda_{max} \rightarrow \text{Mode } \lambda \begin{cases} \lambda : \text{active input} \\ \theta = \theta_{max} \end{cases}$$

$$55 \text{ Case : } \theta = \theta_{max} \wedge \lambda = \lambda_{max} \rightarrow \begin{cases} \text{If } \dot{\lambda} > 0 \rightarrow \text{Mode } \theta \\ \text{If } \dot{\theta} > 0 \rightarrow \text{Mode } \lambda \end{cases}$$

the constrained domain of the complementary inputs is
shown in Fig. 1.

3.2 Extended system model

56 In the next step, the system structure needs to
57 be conditioned. The original IC engine model consists
58 of two first-order differential equations, describing the
59 speed and manifold pressure dynamics. However, in prac-
60 tice, it is preferable to avoid direct discontinuous signals
61 at the inputs of the real actuators of the engine. Therefore,
62 to smooth out the physical inputs, the original system
63 is dynamically extended with the incorporation of inte-
64 grators prior to the actual inputs (*i.e.*, θ and λ become
65 states and their derivatives are the new artificial inputs).
66 The smoothing of the control action contributes to the
67 reduction of the undesirable chattering effect.

Then, the extended system is then fully modelled by:

$$68 \dot{n} = \frac{1}{I}(T_e - T_L) \quad (17)$$

$$69 \dot{p}_i = \frac{RT_i}{V_i}(\dot{m}_{at} - \dot{m}_{ap}) \quad (18)$$

$$70 \dot{\theta} = u_1 \quad (19)$$

$$71 \dot{\lambda} = u_2 \quad (20)$$

72 Considering the complementary operation mode
73 described in the previous subsection, the fourth-order
74 system can now be reduced to two alternating third-order
75 dynamic systems:

In Mode θ then :

$$\begin{aligned} \dot{n} &= \frac{1}{I} (T_e|_{\lambda=\lambda_{max}} - T_L) = f_{n_\theta}(p_i, n, \theta) - \frac{1}{I} T_L \\ \dot{p}_i &= \frac{RT_i}{V_i} (\dot{m}_{at} - \dot{m}_{ap}) = f_p(n, p_i) \\ \dot{\theta} &= u_1; \end{aligned} \quad (21)$$

In Mode λ then :

$$\begin{aligned} \dot{n} &= \frac{1}{I} (T_e|_{\theta=\theta_{max}} - T_L) = f_{n_\lambda}(p_i, n, \lambda) - \frac{1}{I} T_L \\ \dot{p}_i &= f_p(n, p_i) \\ \dot{\lambda} &= u_2. \end{aligned}$$

provided that the state variables θ and λ are forced to lie in the domain shown in Fig. 1.

3.3 SM control algorithm for speed regulation

The final step in the design is to address the fulfillment of the main control objective, that is the regulation of the engine speed at its nominal value. To formalize this objective within the framework of the SM control, a sliding variable is defined to force a first-order linear dynamics for the engine speed error:

$$s = e + \tau \frac{de}{dt} = (n - n_o) + \tau \frac{dn}{dt} \quad (22)$$

where e is the speed error, n_o is the speed reference and τ is a design parameter that defines the stable linear dynamics of n [17,18]. n_o is assumed to be constant, therefore $\frac{de}{dt} = \frac{dn}{dt}$ in (22). Note that the relative degree of s with respect to the inputs of the extended system is one.

With the physical inputs appropriately smoothed, the speed regulation objective embedded in the sliding variable and the fuel minimization objective contemplated in the implementation of the complementary inputs the following SM algorithm is proposed to robustly satisfy both control goals:

In Mode θ then :

$$\begin{aligned} u_1 &= k_1 \text{sign}(s) \rightarrow \theta = \int u_1 dt = \int (k_1 \text{sign}(s)) dt \\ u_2 &= 0 \rightarrow \lambda = \lambda_{max} \end{aligned} \quad (23)$$

In Mode λ then :

$$\begin{aligned} u_1 &= 0 \rightarrow \theta = \theta_{max} \\ u_2 &= k_2 \text{sign}(s) \rightarrow \lambda = \int u_2 dt = \int (k_2 \text{sign}(s)) dt \end{aligned}$$

Before closing this subsection it is worth noting a practical issue. During the implementation, the acceleration of the engine can provide a noisy signal if it is derived from the direct differentiation of the measured engine speed. To counteract this drawback, it could be

computed through a diffeomorphism using speed and manifold pressure, two common measurements in real engines, together with the load torque value, which can be inferred from the electric power demanded by the generator. Another practical option, which is more robust to certain model uncertainties than the former, is to obtain the acceleration from a reduced order observer or different SM estimation methods [19,20]. This issue is being considered in on going research and is not addressed herein.

3.4 Zero dynamics

The system (21) zero dynamics, named Ideal Sliding Dynamics (ISD) in the SM control framework, is the dynamics in \mathbb{R}^2 defined by $s = 0$ and the full dynamics constrained to this manifold. The sliding surface is, by definition an ordinary differential equation. Thus, since the ISD is of dimension 2, we may select the pressure dynamics as the second equation. Hence, the ISD reduces to

$$\dot{n} = \frac{1}{\tau} (n - n_o) \quad (24)$$

$$\dot{p}_i = f_p(n, p_i). \quad (25)$$

Note that the n -dynamics is decoupled; it yields $n(t) = e^{-\frac{t}{\tau}} \cdot (n(0) - n_o) + n_o$. Hence, after some time $n = n_o$ can be assumed and the pressure dynamics results in

$$\dot{p}_i = f_p(n_o, p_i). \quad (26)$$

Its stability can then be straightforwardly determined from the inspection of the plot of $\dot{p}_i = f_p(n_o, p_i)$ for $p_i \in [0, 1]$ (see Fig. 2). Note that $f_p(n_o, p_i)$ cancels at a single point p_i^* within the interval $[0, 1]$, being positive for $p_i \in [0, p_i^*)$ and negative for $p_i \in (p_i^*, 1]$. Hence, p_i^* is asymptotically stable and its basin of attraction is the whole interval $[0, 1]$, amply covering the operation range of p_i .

3.5 Design of controller gains: sliding mode basin of attraction

The proposed procedure for the design discontinuous control gains of the controller is presented in this subsection. Two main features are considered for the calculation of k_1 and k_2 , firstly, to provide the controller with sufficient control authority to deal with a wide range of load torques, namely:

$$T_{L_{min}} = 5 \leq T_L \leq T_{L_{max}} = 54. \quad (27)$$

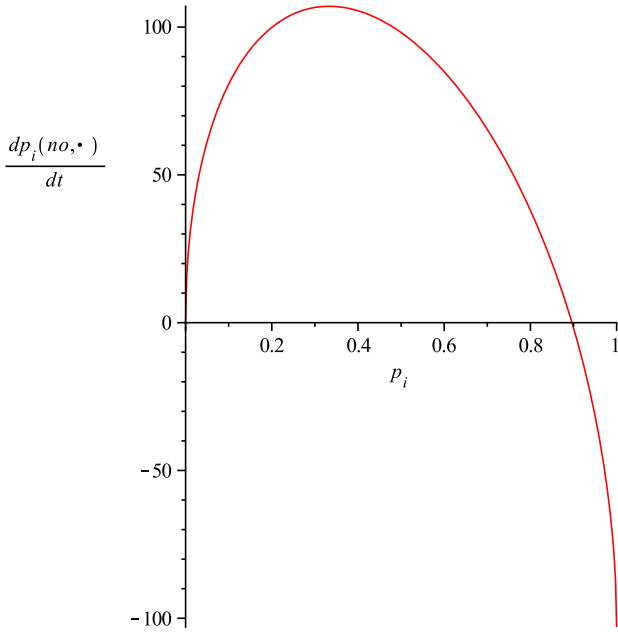


Fig. 2. Graph of $f_p(n_o, p_i)$ for $p_i \in [0, 1]$.

and secondly to guarantee a secure basin of attraction. Taking into account the operation characteristics of the IC engine, the following values were set:

$$s_{min} = -500 \leq s \leq s_{max} = 500. \quad (28)$$

To start the design procedure, the expression of the continuous control action which maintains s invariant at an arbitrary constant value K (hereinafter referred to as u_{eq_K}) is obtained from $\dot{s} = 0$ evaluated at $s = K$:

$$u_{eq_K} = u \Big|_{\substack{s=0 \\ s=K}}. \quad (29)$$

It should be recalled that in *Mode θ* or *Mode λ* , the system is dependent on four variables: three states ($[n \ p_i \ \theta]^T$ or $[n \ p_i \ \lambda]^T$, respectively) and the external load torque T_L . Therefore, taking advantage of the algebraic relationship set by condition $s = K$ in (29), one variable can be eliminated and u_{eq_K} can be expressed as a reduced function of three variables. In this work, the state variables were preferred, so T_L was eliminated, resulting in:

$$\begin{aligned} \text{Mode } \theta &\rightarrow u_{1eq_K}(n, p_i, \theta), \\ \text{Mode } \lambda &\rightarrow u_{2eq_K}(n, p_i, \lambda). \end{aligned} \quad (30)$$

In addition, for the design, a known upper bound for the time derivative of T_L equal to 20 Nm/s was assumed.

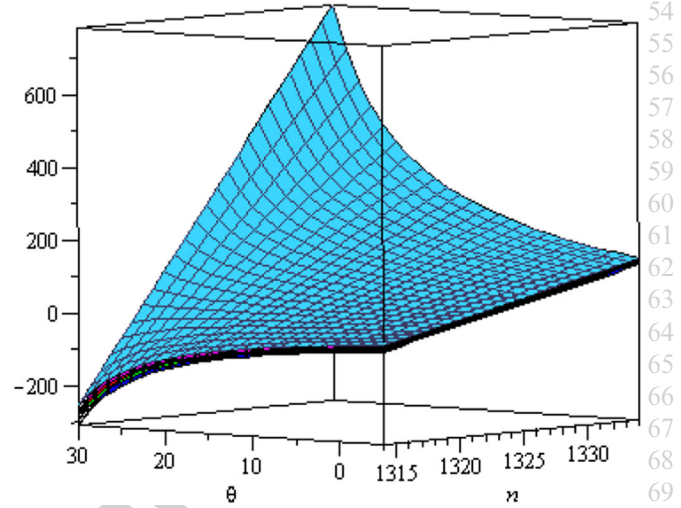


Fig. 3. u_{1eq_K} parametrized in terms of $K = -500, -200, 0, 200, 500$ for a constant $p_i = 0.8963$.

The value for u_{eq_K} in (30) is then computed for the whole state range of operation, considering different values for K , viz.: the external limits ($K = s_{min}$ and $K = s_{max}$) and additional representative internal values ($K = -200, 0, 200$). The former will provide upper and lower bounds for u_{eq_K} which, in turn, are used as the discontinuous control gains to ensure the attraction inside the pre-selected basin.

The operation range of the system under study is:

$$\begin{aligned} n_{min} &= 1294 \leq n \leq n_{max} = 1354 \\ p_{i_{min}} &= 0.8953 \leq p_i \leq p_{i_{max}} = 0.8973 \\ \lambda_{min} &= 0.9 \leq \lambda \leq \lambda_{max} = 1.2 \\ \theta_{min} &= -5 \leq \theta < \theta_{max} = 30 \end{aligned} \quad (31)$$

Figs 3 and 4 illustrate u_{1eq_K} and u_{2eq_K} , respectively, parametrized in terms of $K = -500, -200, 0, 200, 500$ for a single pressure value, in this case the central value $p_i = 0.8963$.

In practice, increasing or decreasing the pressure value results in vertical translations of the u_{1eq_K} and u_{2eq_K} graphs. Hence, computing u_{1eq_K} and u_{2eq_K} for the extreme pressure values, $p_{i_{min}} = 0.8953$ and $p_{i_{max}} = 0.8973$, give the sought-after bounds, namely:

$$\begin{aligned} -800 &\leq u_{1eq_K} \leq 800 \\ -60 &\leq u_{2eq_K} \leq 60 \end{aligned} \quad (32)$$

Consequently, designing $k_1 = 800$ and $k_2 = 60$ guarantees the aforementioned basin of attraction, as

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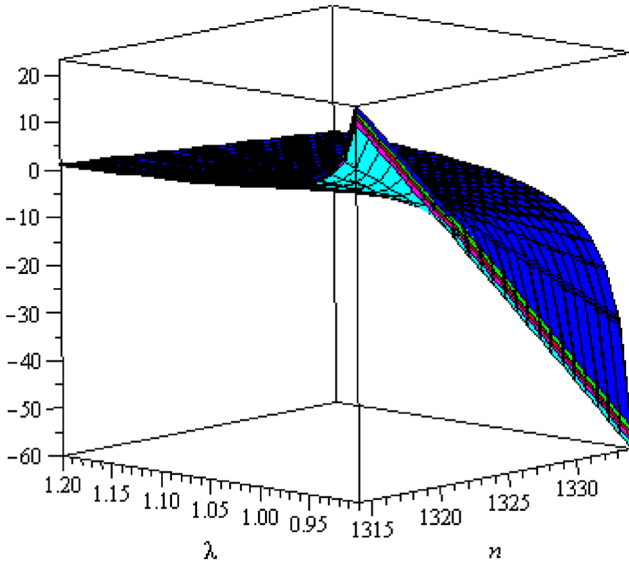


Fig. 4. u_{2eqK} parametrized in terms of $K = -500, -200, 0, 200, 500$ for a constant $p_i = 0.8963$.

well as robustness in relation to the load torque disturbances in (27).

Note that for the design of the gains a 20 Nm/s bound for the load torque derivative was assumed. Therefore, in cases with abrupt time derivatives such as the exacting torque step series presented in the Simulation Results section, the controller cannot ensure insensitivity and the system transiently escapes from the sliding manifold. Nonetheless, it remains inside the attraction basin and hence it robustly returns to $s = 0$.

IV. SIMULATION RESULTS

To determine the feasibility of the proposed SM approach to control a lean-burn IC engine, the performance and robustness of the SM control strategy was assessed through intensive simulation tests. Given that real actuators can saturate, anti wind-up integrators were used in the simulations, and these will be later implemented in the actual engine. In this section, some representative results are presented.

For these tests, the IC engine is subjected to an exacting series of load demand steps (see Fig. 5). In fact, in actual applications the load regime is not so abrupt, but this series was selected to present the controller with an exigent trial.

In addition to observing the controller behaviour it is of interest to establish comparisons, and hence the proposed SM controller with complementary inputs is

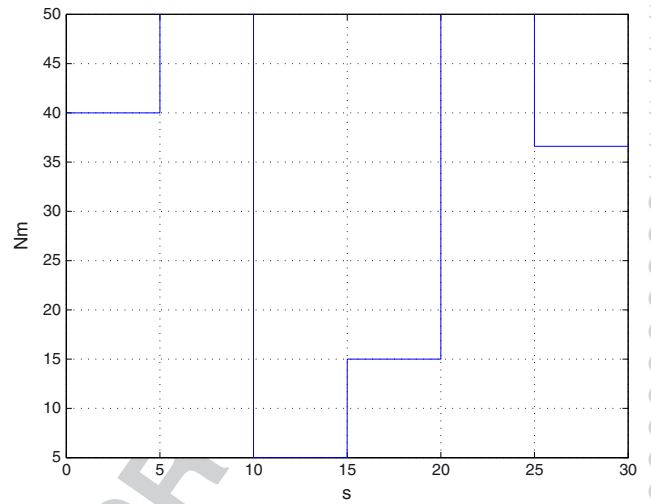


Fig. 5. Load torque ($N - m$).

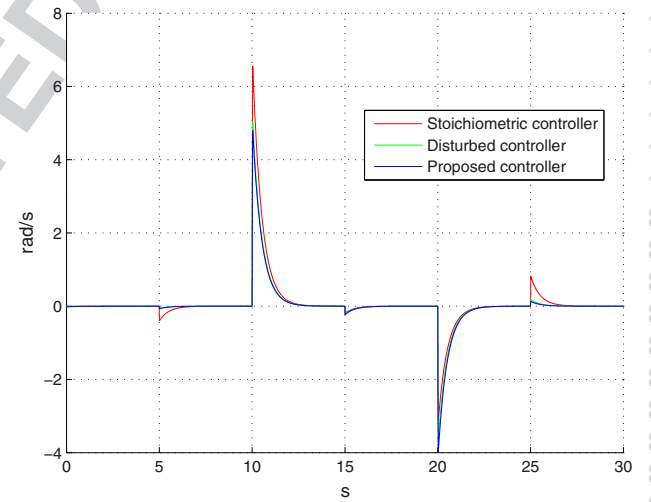


Fig. 6. Engine speed error (rad/s).

contrasted with a stoichiometric SM controller. In the latter, the injector is commanded to maintain the air–fuel mass at a stoichiometric ratio of $a_s = 14.6$ (which implies constant input $\lambda = 1$), while the input θ is controlled by an integrated discontinuous control action ($\theta = \int (k \text{sign}(s)) dt$). The speed error obtained with the two controllers under undisturbed conditions are shown in Fig. 6 (in blue and red). The reference value for the engine angular speed is set at $n_0 = 1324$ rad/s. Note that the values for the error amplitude and the settling times are comparable for both control set-ups.

In order to investigate the proposed controller performance in the presence of lumped disturbances, in a

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subsequent test a 10% error was added to the nominal engine torque. This error term is intended to take into account the effects of model parameter uncertainties and/or unmeasurable perturbations, assuming them to be unknown but bounded. The results obtained satisfactorily confirm the robustness of the controller in relation to disturbances that can be grouped in this form.

Figs 7 and 8 show the time evolution of the spark advance and λ for the stoichiometric SM controller (under nominal conditions of operation) and the proposed SM controller (nominal and disturbed). The complementarity of the proposed inputs can be clearly observed in these figures (see blue and green lines, which correspond to the designed controller).

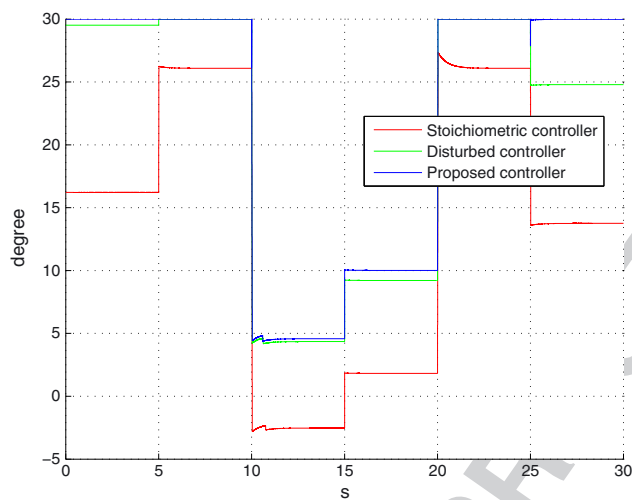


Fig. 7. Spark advance (deg).

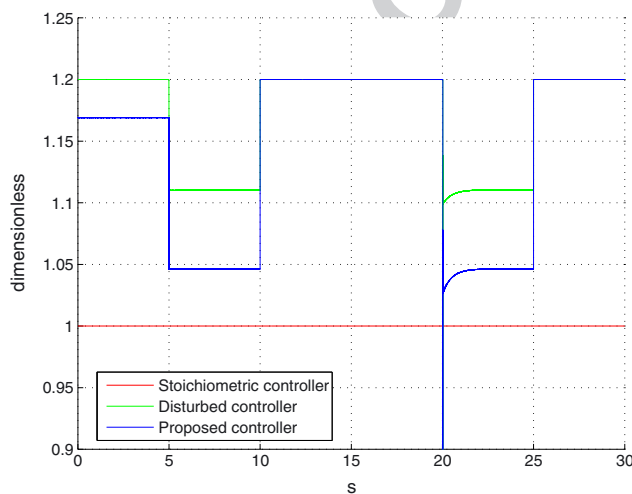


Fig. 8. Lambda.

The success of the proposed complementary inputs strategy in relation to attaining a reduction in fuel consumption can be evaluated through the fuel conversion efficiency, defined as a performance indicator in (11). It can be verified in Fig. 9 that the efficiency of the implementation of the SM complementary inputs is better than that of the engine controlled to operate with a stoichiometric mixture.

In Fig. 10 the trend of normalized NO_x pollutant production is shown. Although the production of pollutants was not an explicit objective of this paper, it is interesting to verify that the proposed controller can also reduce the production of NO_x.

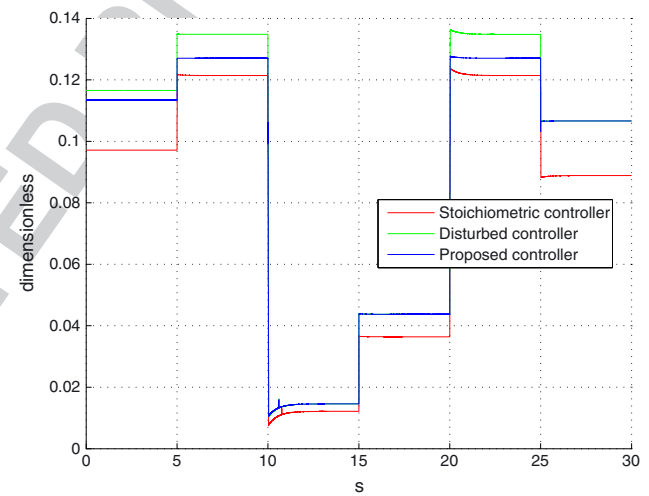


Fig. 9. Fuel conversion efficiency.

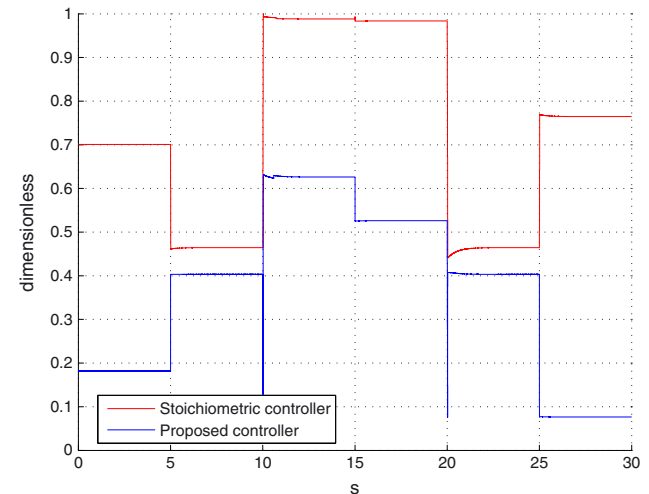


Fig. 10. Normalized NO_x production.

V. CONCLUSIONS AND FUTURE WORK

The results presented in this paper correspond to the first stage of a broader research project aimed at developing a control system for the actual IC engine of a novel hybrid vehicle. The purpose of this initial stage was to design a robust controller for an IC engine based on sliding mode techniques and the implementation of complementary inputs able to deal with lean burn operation. To assess the feasibility of this approach, rather than the actual parameters of the engine under construction, a well-tested simulation benchmark model was used.

For the development of the controller the system was expanded with integrators at the inputs, to provide a smooth control action for the actuators. The designed SM controller successfully fulfilled two simultaneous control objectives: robust regulation of the engine speed at its nominal value and a reduction in the fuel consumption. The former was attained by defining the sliding surface in terms of the speed error. The latter was fulfilled through the implementation of complementary inputs.

Extensive simulation results show that the SM controller is able to deal with a lean-burn engine with wide open throttle by acting only on the air–fuel ratio and spark advance. Moreover, the proposed system topology eliminates an actuator (the throttle) and hence the cost and complexity of the control system are substantially reduced. Finally, it is of particular relevance to highlight that the fuel efficiency attained with the proposed control strategy is better than that obtained with the IC engine operating at a fixed stoichiometric air–fuel ratio.

Encouraged by these promising results, future work will involve the development of an engine model of the actual hybrid power train as well as the design, implementation and experimental testing of the SM based control system in the real engine and further studies on its robustness under exacting conditions.

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 34 tation, Advances in Variable Structure and Sliding Mode
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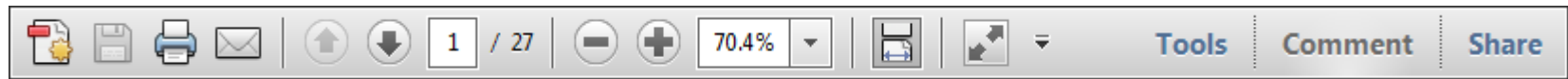
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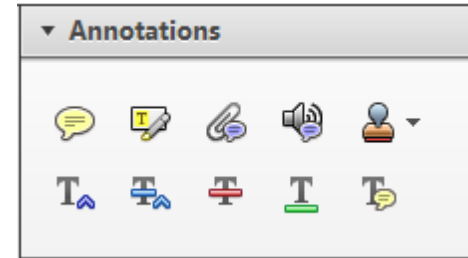
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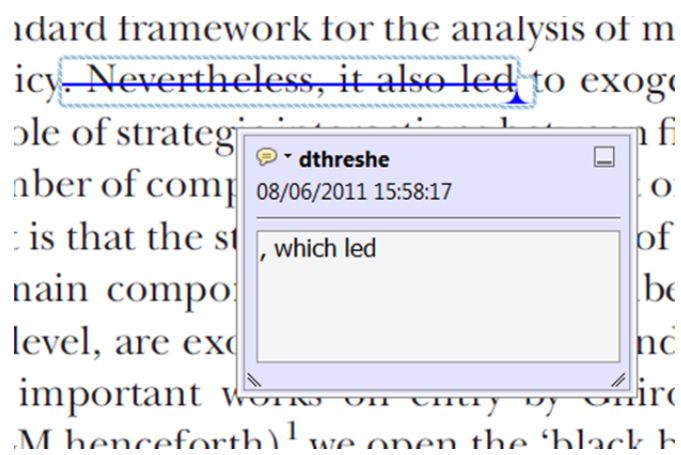
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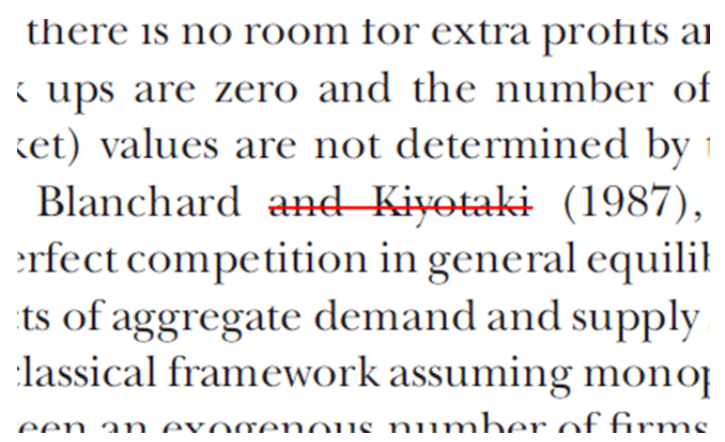
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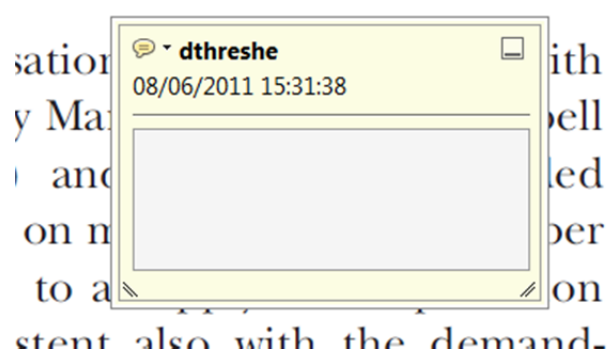
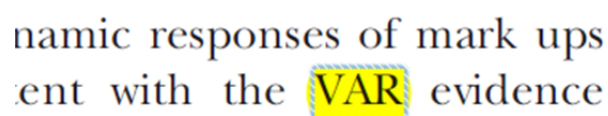
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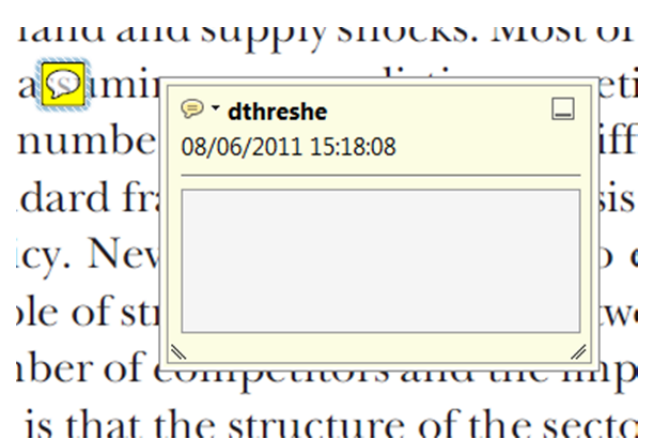
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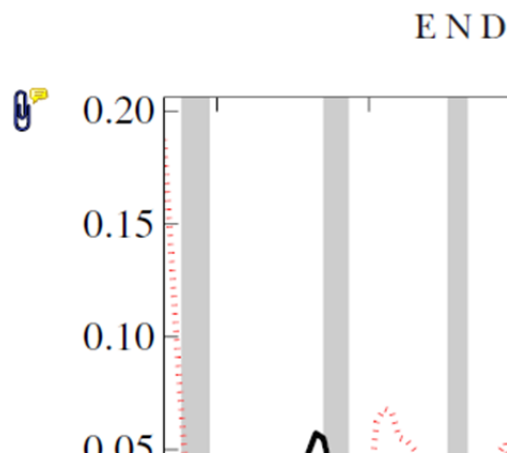
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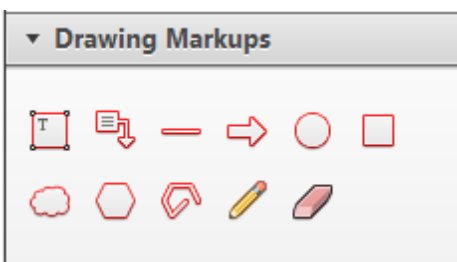


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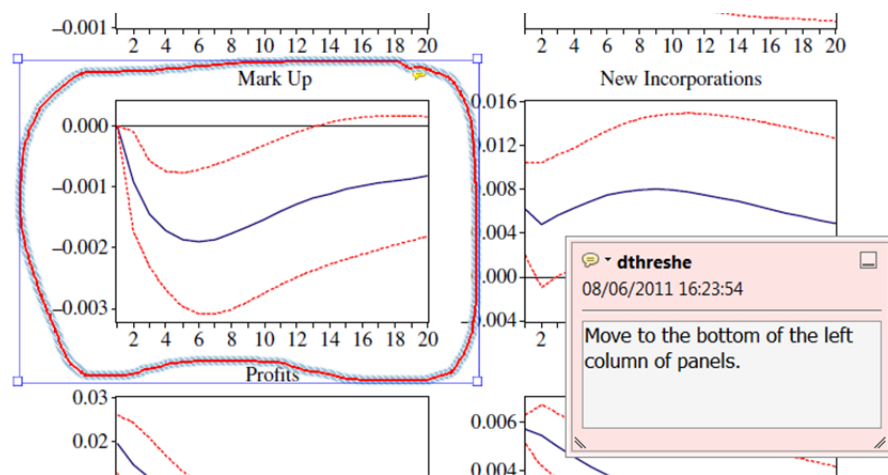


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